

# Low-Power High-Speed Operation of Submicron InP–InGaAs SHBTs at 1 mA

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**Abstract**—Scaling of submicron InP–InGaAs HBTs is investigated for low-power high-speed applications in mixed signal circuits. Device performance for transistors fabricated with a  $0.5\text{-}\mu\text{m}$  emitter width and varying emitter lengths are studied. The  $0.5\text{ }\mu\text{m} \times 2\text{ }\mu\text{m}$  devices yielded excellent low-current RF performance, with an  $f_T = 173\text{ GHz}$  and an  $f_{MAX} = 187\text{ GHz}$  at  $1\text{ mA}$ , the highest values reported for InP-based devices to date.

**Index Terms**—Heterojunction bipolar transistors (HBTs).

## I. INTRODUCTION

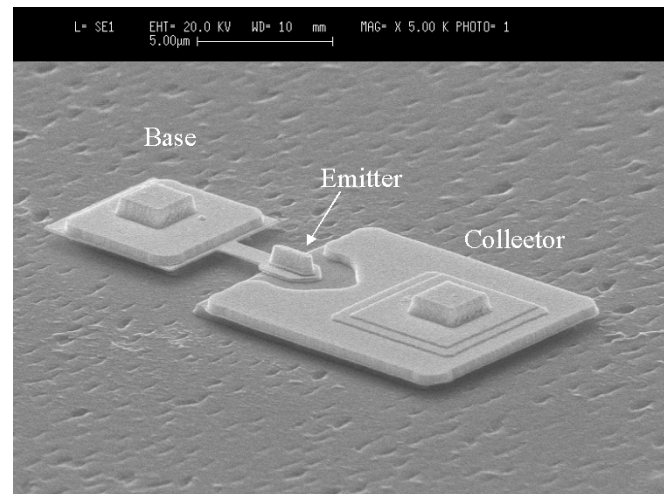
SUBMICRON heterojunction bipolar transistors (HBTs) with high performance at low current levels are crucial to the development of high-speed high-density integrated circuits. Low-power consumption and better thermal management are driving the need for development of deep-submicron devices. This letter will present devices with record RF performance ( $f_T = 173\text{ GHz}$ ,  $f_{MAX} = 187\text{ GHz}$ ) at  $1\text{ mA}$  of collector current for InP-based devices. Comparisons of state-of-the-art low-power devices for both InP-based and SiGe HBTs are made.

## II. LAYER STRUCTURE

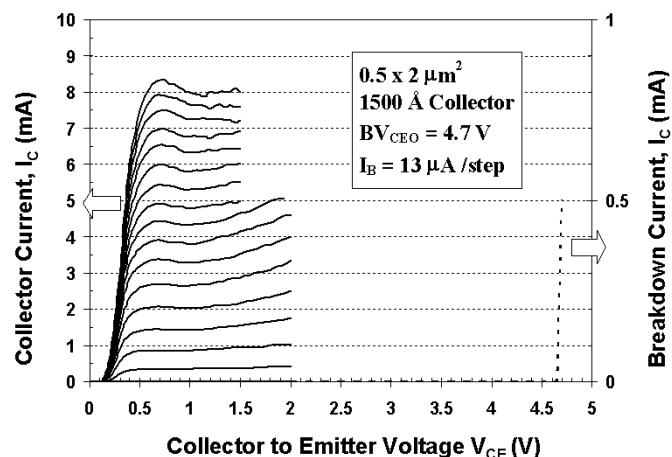
The epitaxial structure is aggressively scaled and utilizes an  $800\text{-}\text{\AA}$  emitter structure, as well as a  $300\text{-}\text{\AA}$  base, Carbon doped to  $5 \times 10^{19}\text{ cm}^{-3}$  and corresponding to a sheet resistance of  $970\text{ }\Omega/\text{square}$ . The base layer is compositionally graded from an Indium mole fraction of  $0.5$  to  $0.53$  to reduce the device transport time. A  $1500\text{-}\text{\AA}$  InGaAs collector, doped at  $1 \times 10^{16}\text{ cm}^{-3}$ , is implemented to further reduce the transit time and increase the current cutoff frequency  $f_T$ . Devices were fabricated using a standard all wet-etch triple-mesa procedure as outlined in [1] and [2].

## III. DC RESULTS

Fig. 1(a) shows a SEM of a fabricated  $0.5 \times 2\text{ }\mu\text{m}^2$  device. The common-emitter family of curves for the same device is shown in Fig. 1(b). Peak RF performance is achieved at a collector current of  $5\text{ mA}$  ( $J_C = 600\text{ kA/cm}^2$ ). The UIUC transistor has dimensions of  $0.5\text{ }\mu\text{m} \times 2\text{ }\mu\text{m}$ , however, lateral etching during processing reduces the physical device area to  $0.83\text{ }\mu\text{m}^2$ , as determined by scanning-electron microscope during fabrication. The current densities reported are calculated taking this area re-



(a)



(b)

Fig. 1. (a) Fabricated  $0.5 \times 2\text{ }\mu\text{m}^2$  device and (b) common emitter output characteristics for a  $0.5 \times 2\text{ }\mu\text{m}^2$  InP–InGaAs SHBT.

duction into consideration. It is also important to note that the minimal self-heating in the device, due to a small emitter area, allows operation above  $1000\text{ kA/cm}^2$  and corresponds to a collector current greater than  $8\text{ mA}$  in Fig. 1. The dc gain  $\beta$  at  $1\text{ mA}$  is  $54$ . Idealities of the base and collector are  $1.36$  and  $1.18$ , respectively. The collector-emitter breakdown  $BV_{CEO}$  is  $4.7\text{ V}$ . An increase in  $BV_{CEO}$  for devices with shorter emitter lengths has been observed. The increase in emitter resistance of the shorter emitter length devices decreases the voltage drop across the base-emitter junction, thereby effectively increasing the breakdown of the device. For example, the  $BV_{CEO}$  is  $3.7\text{ V}$  for  $12\text{ }\mu\text{m}$  emitter length, and  $BV_{CEO}$  is  $4.7\text{ V}$  for  $2\text{ }\mu\text{m}$  length emitter length.

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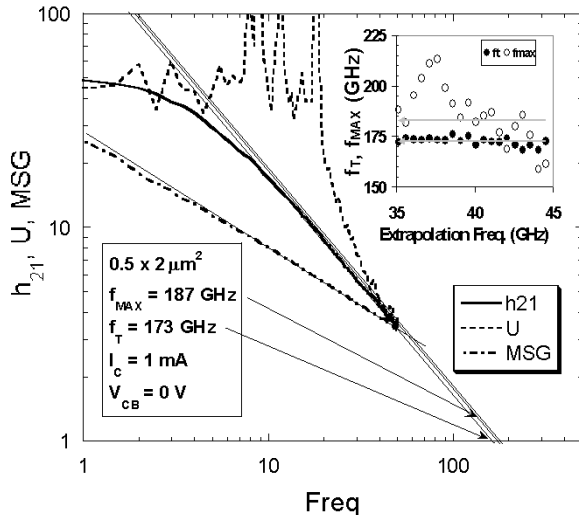


Fig. 2. Extrapolation of  $f_T$  and  $f_{MAX}$  for a  $0.5 \times 2 \mu\text{m}^2$  device at  $I_C = 1$  mA.

#### IV. RF RESULTS

The highest RF performance in the low-current region of operation is achieved in a  $0.5 \mu\text{m} \times 2 \mu\text{m}$  device. At 1 mA of collector current, the device yields an  $f_T$  of 173 GHz with a simultaneous  $f_{MAX}$  of 187 GHz at a  $V_{CE}$  of 0.86 V. At peak performance, an  $f_T$  of 310 GHz and an  $f_{MAX}$  of 275 GHz is achieved at a bias current of 5 mA ( $J_C = 600 \text{ kA/cm}^2$ ). A typical extrapolation of  $h_{21}$ , MSG, and  $U$  is shown in Fig. 2 at 1 mA of collector current. The value of  $f_T$  was taken from averaging  $-20$  dB/decade slope of  $h_{21}$  between 35 and 45 GHz. We observe  $f_T$  remains constant and independent of the frequency where the extrapolation is made, as shown in the inset of Fig. 2. The value of  $f_{MAX}$  was obtained through two methods. The first method followed the same technique as  $f_T$ , with the averaging performed on the Mason's Unilateral Gain parameter  $U$ . The inset of Fig. 2 shows  $f_{MAX}$  varying from 160 to 215 GHz, contributing uncertainty to the final result of 187 GHz. The second method for extracting  $f_{MAX}$  was calculated using the maximum stable gain parameter MSG. For this extraction, the stability factor  $k$  was linearly extrapolated to find the corner frequency where the transistor becomes stable ( $k = 1$  at 62 GHz). The MSG curve in Fig. 2 was extrapolated to this frequency, and beyond this frequency was fitted with a  $-20$  dB/decade slope. This method yields an  $f_{MAX}$  of 190 GHz, in close agreement with the averaging technique described earlier [3].

A plot of  $f_T$  and  $f_{MAX}$  for varying emitter lengths from  $2 \mu\text{m}$  to  $12 \mu\text{m}$  at a fixed width ( $0.5 \mu\text{m}$ ) is shown in Fig. 3. It is observed  $f_T$  shows a weak dependence on emitter length, and remains fairly constant for the larger devices. The highest extrinsic  $f_T$  occurs for a  $0.5 \mu\text{m} \times 16 \mu\text{m}$  device, and is measured at 382 GHz [4]. The large drop in extrinsic  $f_T$  of the  $0.5 \mu\text{m} \times 2 \mu\text{m}$  device, 310 GHz, is resulting from an increase in emitter resistance from the smaller emitter geometry. A much stronger dependence in  $f_{MAX}$  versus emitter length is observed; longer emitter lengths have a reduced power gain cutoff frequency due to higher base-collector capacitances, and more importantly higher base resistances as described in [4].

Table I shows a comparison of small geometry HBT devices and their associated low-current performance. In comparison

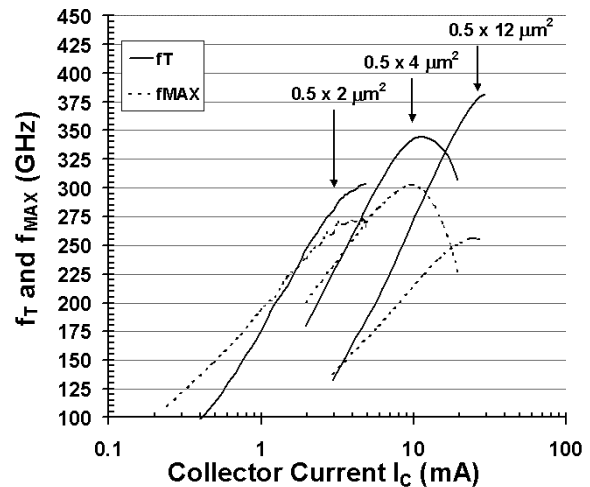


Fig. 3.  $f_T$  and  $f_{MAX}$  versus emitter length for  $0.5 \mu\text{m}$  emitter width.

TABLE I  
PERFORMANCE SUMMARY FOR CURRENT LOW-POWER  
BIPOLAR TECHNOLOGIES

Source	$f_T@1 \text{ mA}$ (GHz)	$f_{MAX}@1 \text{ mA}$ (GHz)	Peak $f_T$ (GHz)	Associated $f_{MAX}$ (GHz)	$W_E \times L_E$ ( $\mu\text{m}^2$ )	$BV_{CEO}$ (V)
HRL InP SHBT [5]	160	Not Reported	162	250	$0.5 \times 1$	$> 2$
IBM SiGe [6]	163	205	270	260	$0.12 \times 2.5$	1.6
IBM SiGe [7]	120	55	210	90	$0.22 \times 5$	1.8
UIUC InP SHBT	145	182	320	310	$0.35 \times 4$	4.4
UIUC InP SHBT	173	187	310	275	$0.5 \times 2$	4.7

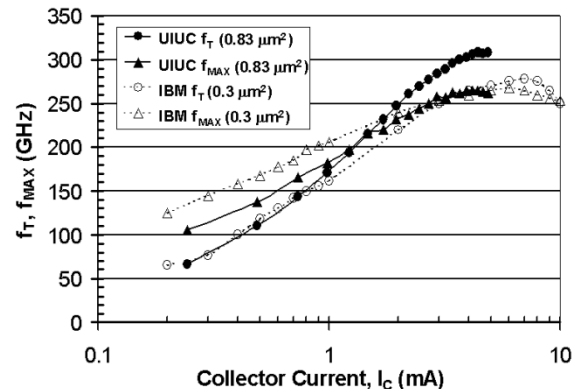


Fig. 4.  $I_C$  versus  $f_T$  and  $f_{MAX}$  for UIUC InP-InGaAs and IBM SiGe devices.

to other InP based HBTs [5], the UIUC devices achieve the highest reported value of  $f_T$  (173 GHz) and  $f_{MAX}$  (187 GHz) at 1 mA. Fig. 4 shows collector current versus cutoff frequency for the  $0.83 \mu\text{m}^2$  UIUC device and the most recent  $0.3 \mu\text{m}^2$  IBM SiGe transistor [6]. The UIUC SHBT displays comparable performance to the IBM SiGe device, which is nearly three times smaller in area and running at over four times the current density of the UIUC device. For a SiGe device with a comparable emitter area of  $1.1 \mu\text{m}^2$  [7], the UIUC devices exhibit far superior  $f_T$  and  $f_{MAX}$ , both at 1 mA and at peak performance. These results indicate that as a more mature submicron processing technology evolves, InP is the most suitable candidate

for supporting low-power high-speed applications as this material system has lower turn-on voltages, higher breakdowns, and higher speeds when compared with silicon-germanium.

## V. CONCLUSION

SHBT transistors exhibiting excellent RF and dc performance at low current levels were fabricated using a vertically and laterally scaled submicron process. Reduction of the emitter length for a fixed emitter width results in a minor decrease in  $f_T$  and a stronger increase in  $f_{MAX}$ . Such devices exhibit sufficient RF performance to sustain 40 Gb/s performance with currents under 1 mA. Further lateral scaling of the emitter width should allow significantly higher performance.

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